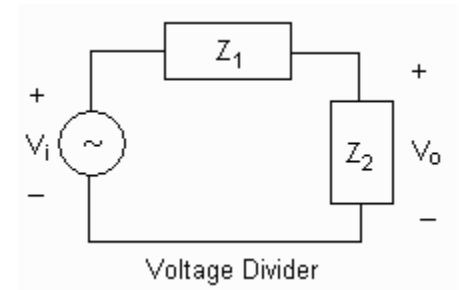


## Voltage Divider

The basic voltage divider on the left will produce  $V_o = V_i * Z_2 / (Z_1 + Z_2)$ , where  $Z_1$  and  $Z_2$  may be complex.



## Frequency Response

Capacitors (and inductors) change characteristic impedance as the frequency changes. In this lab, we will use a computer to capture the frequency response of RC filters by applying an AC frequency sweep to each filter input. During this sweep, the constant output voltage of a function generator is applied to the input of a circuit while the frequency is varied. The output voltage of the circuit is measured at each frequency step and the circuit response is plotted as  $V_o$  vs frequency, with a horizontal logarithmic frequency scale and either a linear voltage or a logarithmic dB (decibel) vertical scale.

## The Decibel

dB is a logarithmic representation of gain or loss used to study or compare circuit responses. The most common form of dB uses a power ratio where  $\text{dB} = 10 \log (P_2/P_1)$ . Circuit gains or losses are measured against a reference ( $P_1$ ), usually the input power or voltage. Decibels are used extensively in electronic system calculations because the dB gains and losses of the response of antennas, mixers, filters, amplifiers, transmission paths, etc can be added together to obtain the total system gain or loss. The reference power may be any reference level – commonly used references are  $P_{in}$ ,  $P_{max}$ , 1W (dBW), 1 mW (dBm), or 1V (dBV).

Using dB,  $1/2$  of the reference power (i.e.  $P_{meas}/P_{ref} = 1/2$ ) would be  $10 \log (1/2) = -3\text{dB}$   
some handy dB values that are frequently seen:

2x the reference power (a power gain of 2) would be  $10 \log (2) = 3 \text{ dB}$

4x the reference power = 6dB

10x the reference power = 10dB

100x the reference power = 20 dB.

$1/2$  the reference power = -3dB

$1/4$  the reference power = -6dB

$1/10$  the reference power = -10dB

$1/100$  the reference power = -20dB

Most frequency response plots will use the dB scale, where gain in dB =  $10 \log (P_2/P_1)$ . We do not measure power in this lab, but  $P = V_2^2/R$ . We can measure two voltages and if the same R is used for both voltage measurements, the R values will cancel and  $\text{dB} = 20 \log (V_2/V_1)$ . The reference voltage  $V_1$  can be input voltage, max output voltage, or some other value. When  $V_2 / V_1 = 1/\sqrt{2}$ , the output power of  $V_2$  is  $1/2$  of the power of  $V_1$  and 3 db below the power of  $V_1$  ( $20 \log 1/\sqrt{2} = -3$ ).

## Capacitor Impedance

The impedance of a capacitor is  $Z_C = 1/j\omega C$ . A capacitor's impedance will be  $\infty\Omega$  when  $\omega=0$ , and the capacitor's impedance will be  $0\Omega$  when  $\omega=\infty$ . When capacitors are combined with resistors, the circuit's impedance and output voltage will be a function of frequency. There will be frequency (called the cutoff frequency) where  $|P_o \text{ at cutoff}| = \frac{1}{2}|P_o \text{ max}|$  or  $|V_o \text{ at cutoff}| = \sqrt{1/2} |V_o \text{ max}|$  ( $\sqrt{1/2} = 0.707$ ).

## Calculating The Cutoff Frequency Of A Circuit

Consider a simple low pass circuit where  $V_o = V_i * X / (Y + j\omega Z)$ .

$V_o \text{ maximum} = V_i * (X / Y)$  when  $j\omega Z = 0$ . Note: X and Y must be real and not functions of  $\omega$ .

Now, by definition the cutoff frequency occurs at half power (when  $V_o = 0.707 V_{\text{max}}$ .) and we need to find where, in terms of  $\omega$ , the magnitude of the output voltage produces this "half power".

So:  $|V_o| = |V_i * X / (Y + j\omega Z)|$  and we want  $|V_o|$  to equal  $0.707 V_{\text{max}} = 0.707 V_i * (X/Y)$  at the cutoff freq.

Note: *this is NOT the same as  $|V_o| = 0.707$  !!!*

Squaring the magnitude of both sides we get

$X^2 / (Y^2 + \omega^2 Z^2) = (\frac{1}{2}) (X^2 / Y^2)$  and the X terms cancel to give  $1 / (Y^2 + \omega^2 Z^2) = 1 / 2Y^2$

Cross multiply and collect terms to get  $Y^2 = \omega^2 Z^2$  and now  $Y = \omega Z$

where Y was the real part of the denominator and  $\omega Z$  was the imaginary part of the denominator.

Solve for  $\omega$ , then  $f = \omega / 2\pi$ .

The high pass equation will have the form  $V_o = V_i * X / (Y + 1/j\omega Z)$  and the process is similar.

## High Pass Filter Response

The high pass filter allows high frequencies to pass through to the circuit output but blocks lower frequencies.

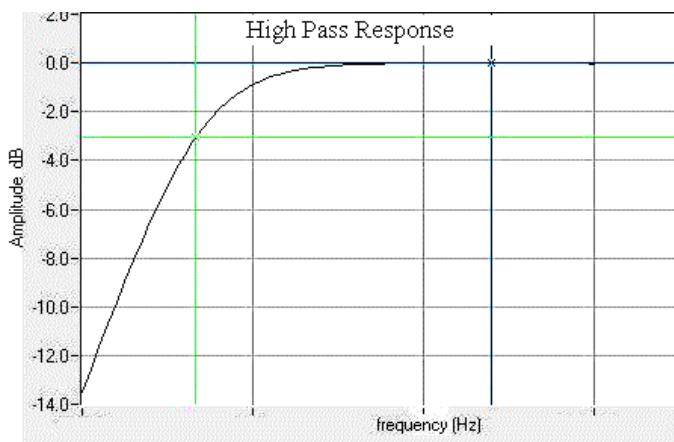


Figure 1

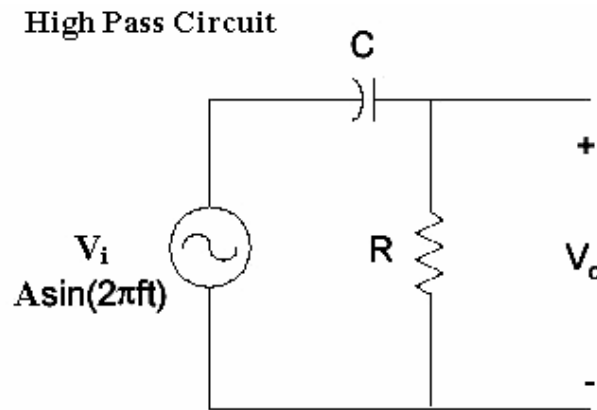


Figure 2

In the high pass response of fig 1 for circuit of fig 2, the cutoff frequency is marked by the green cursor where the amplitude is -3 dB. You can see that frequencies above the cutoff are passed because C is effectively a short circuit, and frequencies below the cutoff are increasingly attenuated as C becomes an open circuit. The response of this circuit is  $V_o = V_i R / [R + 1/j\omega C]$ , and the maximum output voltage ( $V_o = V_i$ ) occurs when  $1/j\omega C = 0$ . The passband is defined as all frequencies above  $\omega_{\text{low}}$ , and the passband gain is defined as the maximum filter amplitude in dB (0 dB in this case).

What happens if we add a resistor  $R_S$  in series with C?

C and the  $R_S$  become a complex impedance ( $Z = R_S + 1/j\omega C$ ) in series with R.

By voltage division,  $V_o = V_i R / (R + Z)$ . With a little algebra the real part of the denominator becomes  $R_S + R$ , the imaginary part becomes  $1/j\omega C$ , and we have  $V_o = V_i R / [(R_S + R) + 1/j\omega C]$

The cutoff frequency will occur when the real part and imaginary part are equal.  
 In the passband  $V_o = V_i R / [(R_s + R)]$  because  $1/j\omega C$  goes to 0 as  $\omega$  becomes large.

## Low Pass Filter Response

The low pass filter allows low frequencies to pass through to the circuit output but blocks higher frequencies.

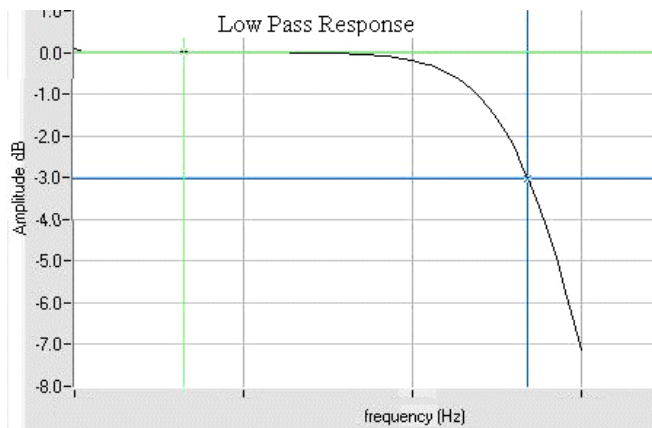


Figure 3

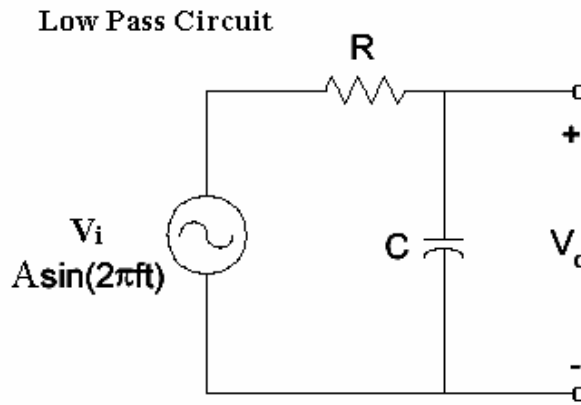


Figure 4

In the low pass response of fig 3 for circuit of fig 4, the cutoff frequency is marked by the blue cursor where the amplitude is -3 dB. You can see that frequencies below the cutoff are passed because  $C$  is effectively an open circuit, and frequencies above the cutoff are increasingly attenuated as  $C$  becomes a short circuit. The response of this circuit is  $V_o = V_i / (1 + j\omega RC)$ , and the cutoff frequency is  $\omega_{high} = 1 / RC$ . The passband is defined as all frequencies below  $\omega_{high}$ , and the passband gain is defined as the maximum filter amplitude in dB (0 dB in this case).

If we add a load resistor  $R_p$  in parallel with  $C$ , the parallel combination of  $R_p$  and  $C$  becomes  $Z = R_p / (1 + j\omega R_p C)$ . By voltage division,  $V_o = V_i Z / (R + Z)$ . With a little algebra the real part of the denominator becomes  $R_p + R$ , the imaginary part becomes  $j\omega R_p R C$ , and we have  $V_o = V_i R_p / [(R_p + R) + (j\omega R_p R C)]$ . The cutoff frequency will occur when the real part and imaginary part are equal. Hint  $R_p R / (R_p + R) = R_p \parallel R$ . In the passband  $V_o = V_i R_p / [(R_p + R)]$  because  $j\omega C$  goes to 0 as  $\omega$  becomes very small.

## Band Pass Filter Response

A band pass filter results from the combination of low pass and high pass circuits, and can be made tight enough to pass a narrow frequency range, such as a single radio station.

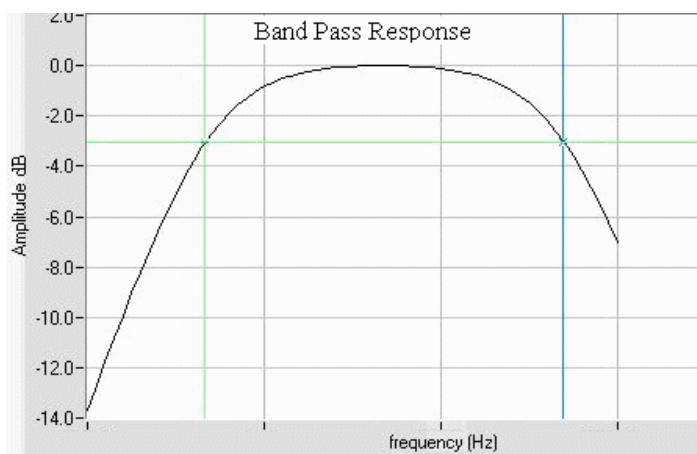


Figure 5

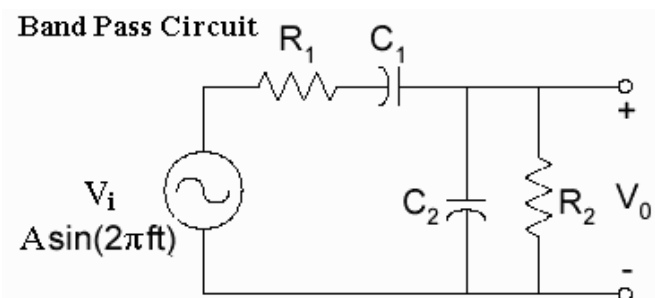


Figure 6

By combining the circuits of figure 2 and figure 4, we obtain the bandpass filter of figure 6.

For this bandpass filter, a highpass response ( $R_1, C_1, R_2$ ) and a lowpass response ( $R_1, C_2, R_2$ ) combine to produce  $V_o = V_i R_2 / \{R_1 + R_2 + j\omega R_1 R_2 C_2 + 1/j\omega C_1\}$

If  $C_1 \gg C_2$ , the output response shown in figure 5 has two cutoff frequencies  $\omega_{low} = 1/ C_1(R_1+R_2)$  and  $\omega_{high} = 1/ C_2(R_1||R_2)$ . In this circuit, both  $C_1$  and  $C_2$  are open at very low frequencies and there is no output voltage from the circuit. As frequency increases,  $C_1$  begins to conduct and allow some of the input signal through the circuit. At frequencies above  $\omega_{low}$ ,  $C_1$  is effectively a short circuit the filter produces its maximum output. As frequency increases to  $\omega_{high}$ ,  $C_2$  begins to conduct, reducing the filter output voltage. As the frequency increases above  $\omega_{high}$ ,  $C_2$  is effectively a short circuit and the filter output voltage steadily decreases (NOTE:  $C_1$  is still a short circuit!). The passband is defined as the frequencies between  $\omega_{low}$  and  $\omega_{high}$ , and the passband gain is defined as the maximum filter amplitude in dB (0 dB in this case). The passband voltage gain  $V_{oMax}/V_i = R_2 / \{R_1 + R_2\}$  because both capacitors are gone ( $C_2$  open and  $C_1$  short circuited) in the passband.

## Measuring the Cutoff Frequency of a Circuit

The cutoff frequency of a circuit is defined as the frequency where output power is reduced to half of the maximum output power. For filters, the passband output voltage or power may be used as the reference (0 dB), placing the filter cutoff at -3dB. If the input voltage is used as the reference, the passband gain or loss =  $20 \text{ Log } V_o / V_i$ . The cutoff frequency will occur where the output drops to 3 dB below the maximum.

Because  $P = V^2/R$ , the equation for dB in terms of voltage becomes  $\text{dB} = 20 \text{ log } (V_2/V_1)$  when the resistor  $R$  is the same for both voltage measurements. The square relationship between power and voltage means that for  $1/2$  power,  $V_{\text{half power}} = .707 V_{\text{reference}}$ . Our lab measurements are voltages, so we will be using  $\text{dB} = 10 \text{ log } (V_2^2/V_1^2) = 20 \text{ log}(V_2/V_1)$ .

Your measurements may differ from the calculated cutoff frequencies due to resistor and capacitor tolerance, the stray capacitance in the protoboard and probes, and interpolation between captured data points with the cursors.

## Modeling a filter in Pspice

Build the circuit in Pspice schematic (as you did for previous Labs) and save your file.

Draw > Get New Part > c > place, position, set the value, and wire the capacitor

Draw > Get New Part > VAC > place, position, set the value AC Mag = 3.54, and wire the AC source

Note: this sets the amplitude to 3.54 Vp, but, because we are not doing a transient analysis where the units are obvious and measurable, we will treat the voltages as RMS for simplicity.

Set up Analysis for AC sweep

Analysis > setup > uncheck all boxes, check AC Sweep >

Click AC Sweep button > set

AC Sweep Type = decade

Pts/Decade = 40

Start Freq = 1

End Freq = 10k

OK > Close > Run the sweep and debug if there are errors.

On the simulation results: Add a dB trace for  $V_{out}$  referenced to  $V_{in}$ ..

[ Trace > Add Trace (or click the add trace button)

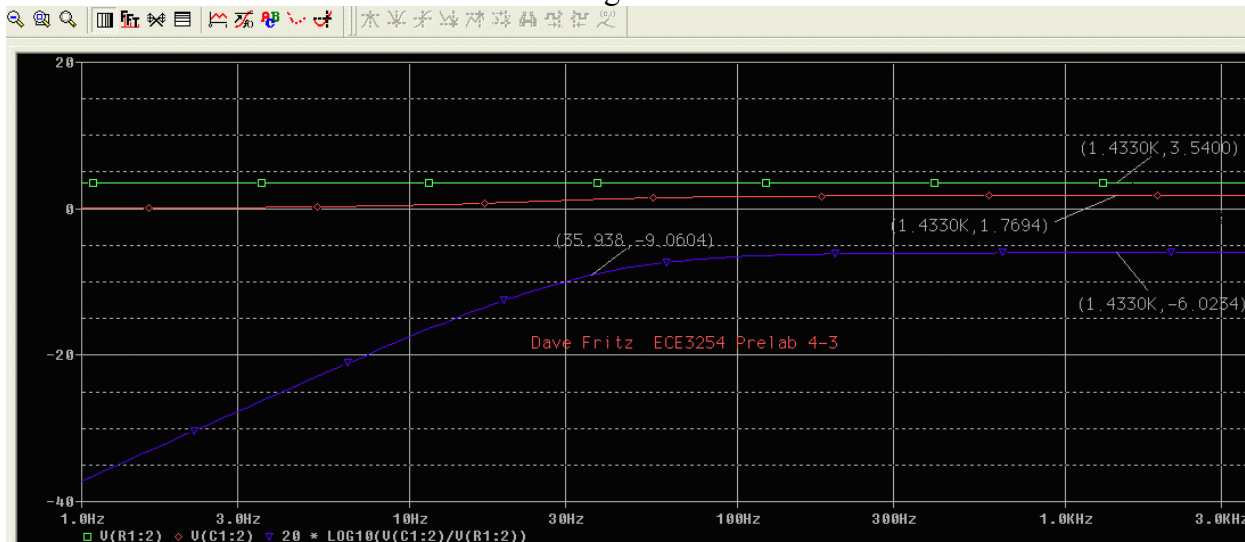
> In Functions or Macros, select Analog Operators and Functions > DB()

> Under Simulation Output Variables, select  $V(C1:2)$ , type /, then select  $V(R1:2)$  > OK ]

\*Note: Instead of using the drop downs, you may add the trace and manually enter the formula  $20 \cdot \text{LOG}_{10}(V(C1:2)/V(R1:2))$  for the new trace\*

\*Note: select the voltages that are correct for your schematic topology.\*

You should see three traces that look something like this:



When you have the dB trace completed, the cursor values placed, and your name on the simulation, print the page.

Adjust the components, wiring, and values for each filter type. For each version of the filter, make sure that you save your schematic.

#### Tip to save time:

- Add C2 to the first model, edit the information box at the bottom, and save the schematic as Prelab 4-9 (or something you can remember). Edit the simulation stop frequency and run the simulation of question #9 at this point. Save the schematic. Print the schematic and simulation.
- Then remove C1, replace C1 with a wire, and save the schematic as Prelab 4-7 (or something you can remember). Edit the simulation start and stop frequencies, and run the simulation for question #7. Save the schematic. Print the schematic and simulation.