

# ECE 3254 PreLab 10 notes

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## Op Amp Active Filters

Op Amp active filters are widely used to limit the bandwidth of signals. Active high pass and low pass elements may easily be combined to form band pass and band stop (notch) filters.

### Inverting Low Pass Filter

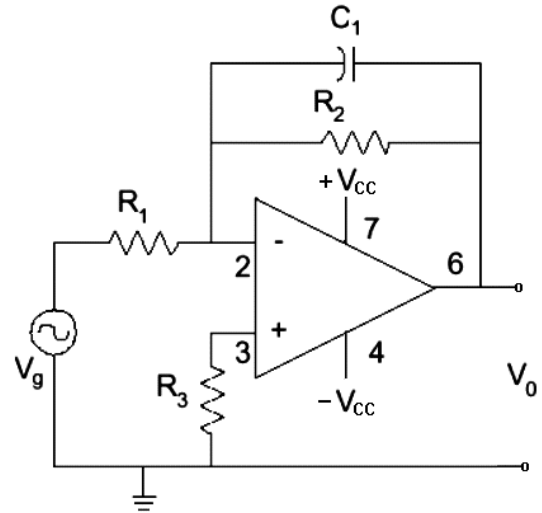
The Op Amp ideal characteristics mean that  $I_{R3} = 0$ ,  $V_{pin2} = V_{pin3} = 0$ , so  $V_{pin2} = 0$ .

From Lab 7,  $V_o = -V_g (Z_2/R_1)$  where  $Z_2$  is the complex parallel equivalent of  $R_2$  and  $C_1$ . Remember that  $Z_C = 1/j\omega C$

$$Z_2 = 1/(1/R_2 + 1/Z_{C1}) = R_2 / (1 + j\omega R_2 C_1)$$

$V_o / V_g =$  you derive from the information above

Note: This circuit is identical to the integrator from lab 9. When operated well above the cutoff frequency, this low pass filter becomes an integrator.



What happens at  $\omega = 0$  (passband response)? The capacitor is an open circuit,  $j\omega R_2 C_1 = 0$ , and  $V_{max} = V_o$  in the passband /  $V_g = -R_2 / R_1$  as in Lab 7. Output phase is  $180^\circ$  due to the inverting Amplifier configuration.

Above the cutoff frequency, the denominator is dominated by the capacitor and the output phase will be adjusted by  $+90^\circ$  to an ultimate phase of  $90^\circ$

How do you find the cutoff freq? You are solving for the frequency where  $|V_o| = .707 |V_{max}|$ . This is NOT where  $|V_o| = 0.707$ . The correct result should be an equation where the  $V_o/V_g$  denominator real part equals the  $V_o/V_g$  denominator imaginary part, and you solve for  $\omega$  as you did in Lab 3.

### Non-Inverting High Pass Filter

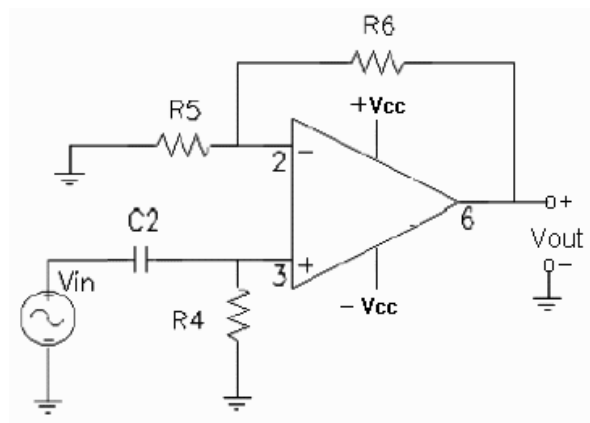
The Op Amp ideal characteristics mean that  $I_{pin3} = 0$  and  $V_{pin2} = V_{pin3}$

$$\begin{aligned}
 \text{By voltage division: } V_{pin3} &= V_{in} R_4 / (R_4 + Z_{C2}) \\
 &= V_{in} R_4 / (R_4 + 1/j\omega C_2)
 \end{aligned}$$

$$\text{From Lab 7, } V_o = V_{pin3} (1 + R_6/R_5)$$

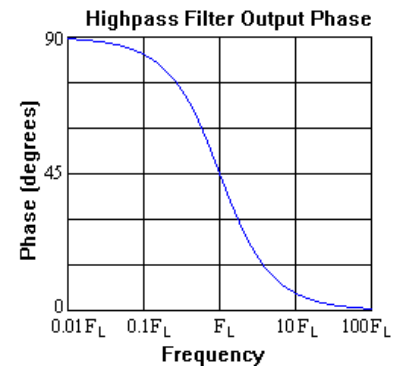
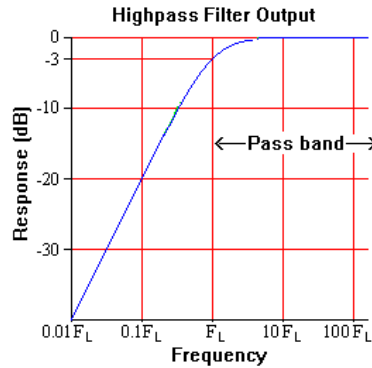
$$\text{So } V_o = V_{in} (1 + R_6/R_5) [R_4 / (R_4 + 1/j\omega C_2)]$$

What happens when  $\omega = \infty$  (passband response)? The capacitor short circuits and  $V_o/V_{in} = 1 + R_6/R_5$



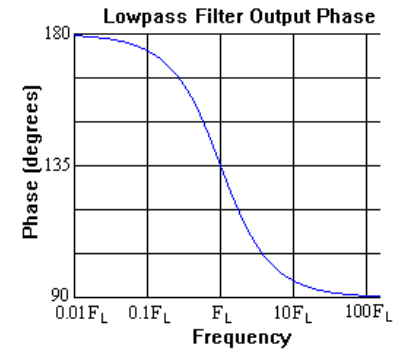
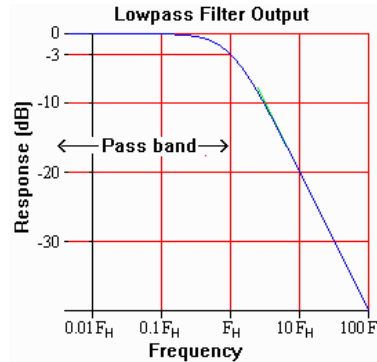
## Active High pass Filter

The passband extends from the cutoff frequency ( $F_L$ ) to  $\infty$ . At very low frequencies the capacitor creates a  $+90^\circ$  phase shift in the output signal. At the cutoff frequency the phase angle is  $45^\circ$  (where real = imaginary). At very high frequencies the capacitor produces no phase shift because it acts as a short circuit.



## Active Low pass Filter

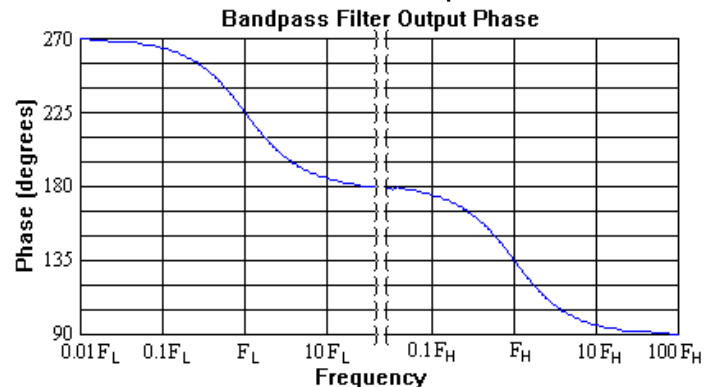
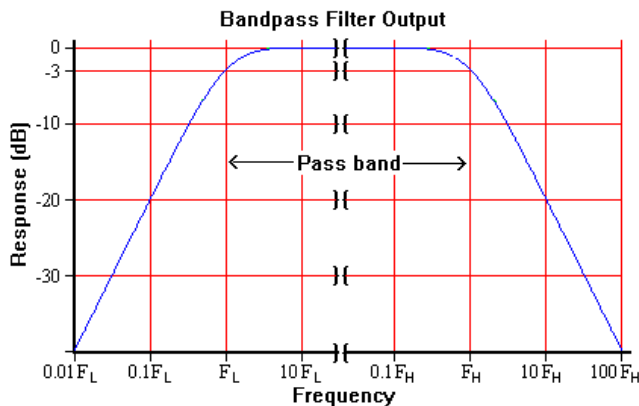
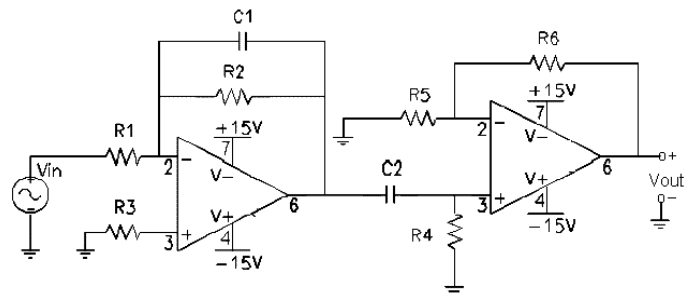
The passband extends from DC ( $F = 0$ ) to the cutoff frequency ( $F_H$ ). At very low frequencies the capacitor is an open circuit and does not affect the phase of the transfer function and the inverting amplifier configuration produces a  $180^\circ$  phase shift in the output. The phase angle is shifted by  $-45^\circ$  (where real = imaginary) at the cutoff frequency, resulting in an output phase of  $135^\circ$ .



At very high frequencies the capacitor adds  $-90^\circ$  to the phase, shifting the output from  $180^\circ$  to  $90^\circ$ .

## Band pass Filter

The high pass filter and the low pass filter can be cascaded by connecting them in series to create a band pass response.



The pass band extends from  $F_L$  to  $F_H$ , and the dB response and phase angles of the two filters add together. At very low frequencies the output phase is  $180+90=270^\circ$  (Note:  $270^\circ$  can also be measured as  $-90^\circ$ ). The output phase passes through  $225^\circ$  (or  $-135^\circ$ ) at  $F_L$ , then through  $180^\circ$  (or  $-180^\circ$ ) in the center of the passband, then through  $135^\circ$  (or  $-225^\circ$ ) at  $F_H$ , and then to  $90^\circ$  (or  $-270^\circ$ ) at very high frequencies.

**Pspice simulation:** Modify the Schematics from Lab 9 to build one filter. Add a second OpAmp to complete the Bandpass filter. Review the Prelab 4 notes for tips on setting up the AC Sweep simulation.